

# UNIT 4

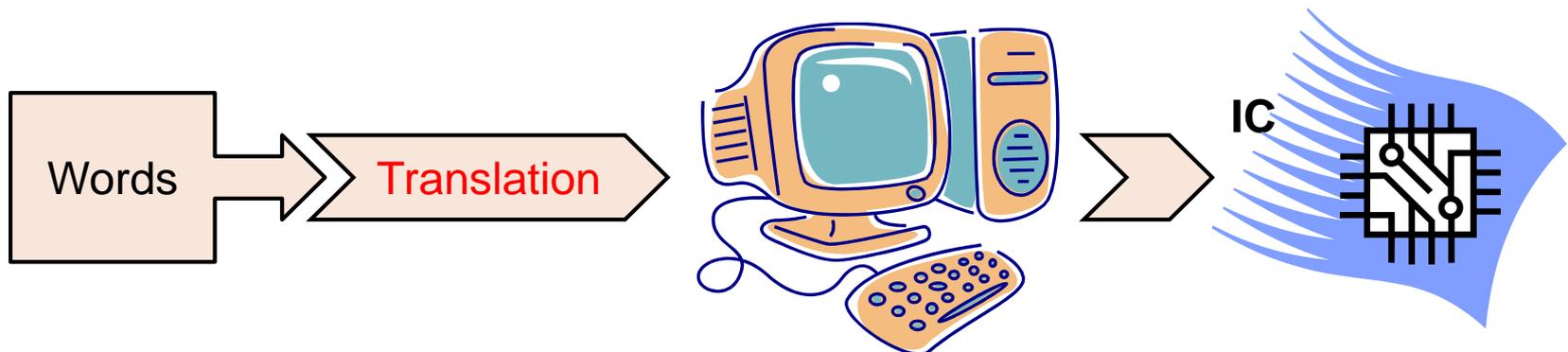
## MINTERM AND MAXTERM EXPANSIONS



Fall 2021

# Minterm and Maxterm Expansions

- **Contents**
  - ▣ Conversion of English sentences to Boolean equations
  - ▣ Combinational logic design using a truth table
  - ▣ Minterm and maxterm expansions
  - ▣ Incompletely specified functions
  - ▣ Examples of truth table construction
  - ▣ Design of binary adders
- **Reading**
  - ▣ Unit 4

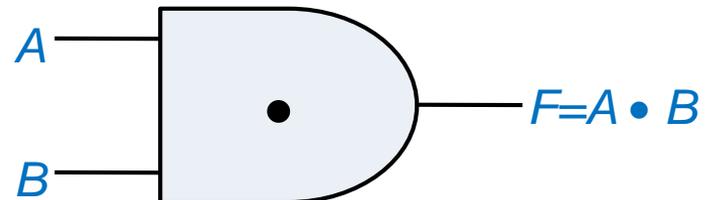


# Objectives

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- **Design a combinational logic circuit starting with a **word description** of the desired circuit behavior**
- **Steps:**
  1. Translate the word description into a switching function
    - Boolean expression or truth table
  2. Simplify the function
  3. Realize it using available logic gates
- **e.g., Mary watches TV if it is Monday night and she has finished her homework**
  - $\Rightarrow$  Translate...
  - **Mary watches TV** ( $F$ ) if
  - **It is Monday night** ( $A$ )
  - **And**
  - **She has finished her homework** ( $B$ )
  - $\Rightarrow F = A \bullet B$



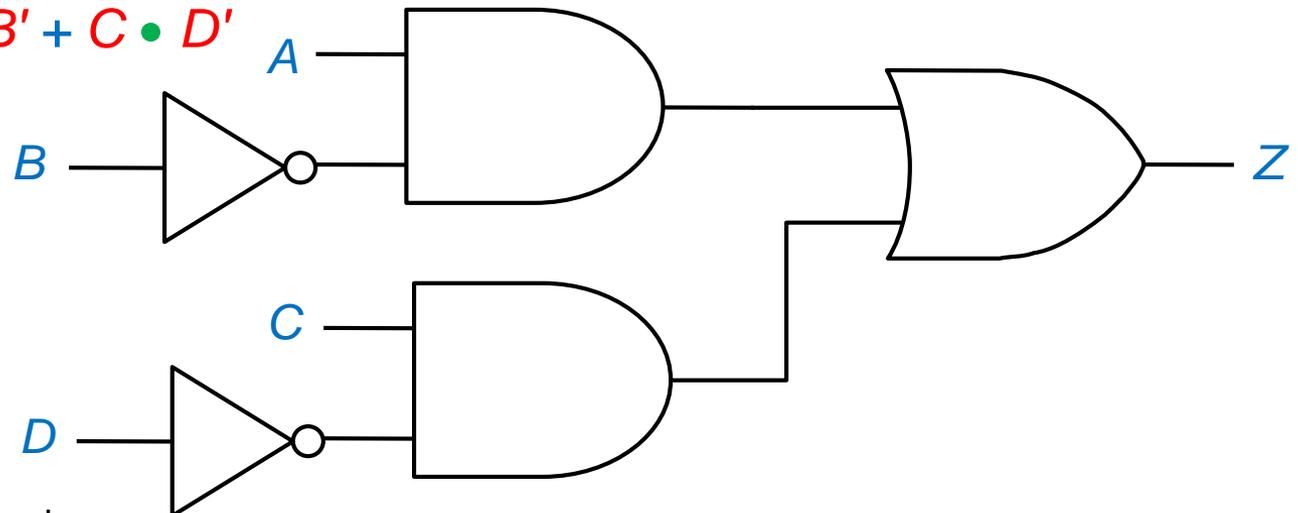
# Example: Designing an Alarm



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- The alarm will ring iff the alarm switch is turned on and the door is not closed, or it is after 6 P.M. and the window is not closed
  - $\Rightarrow$  Translate...
  - The alarm will ring ( $Z$ ) iff
  - The alarm switch is on ( $A$ ) and the door is NOT closed ( $B'$ )
  - Or
  - It is after 6 P.M. ( $C$ ) and the window is NOT closed ( $D'$ )
  - $\Rightarrow Z = A \cdot B' + C \cdot D'$



Minterm & maxterm expansions

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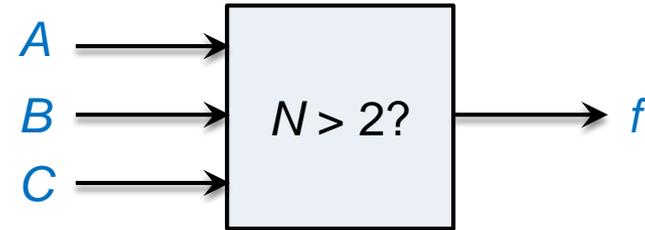
## Logic Design using a Truth Table

**A truth table helps translation**

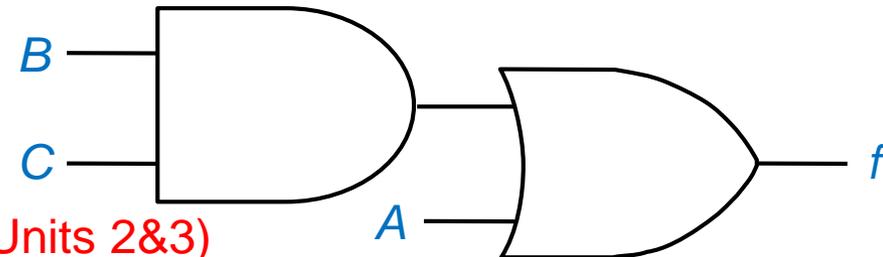
# Threshold Detector (1/2)

- Design a detector that outputs 1 when input is greater than 2
  - ▣ Inputs  $(A, B, C)_2$  represent an unsigned binary integer  $N$ ;  
if  $N = (A, B, C)_2 \geq 3 = 011_2$ , output  $f = 1$ ; otherwise  $f = 0$

	A	B	C	f	f'
0	0	0	0	0	1
1	0	0	1	0	1
2	0	1	0	0	1
3	0	1	1	1	0
4	1	0	0	1	0
5	1	0	1	1	0
6	1	1	0	1	0
7	1	1	1	1	0



Show the condition to make output == 1

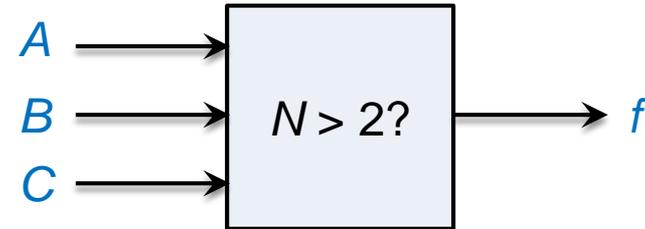


$$\begin{aligned}
 f &= A'BC + AB'C' + AB'C + ABC' + ABC \text{ (SOP)} \\
 &= \underline{A'BC + ABC} + AB'C' + AB'C + ABC' + ABC \\
 &= A + BC \text{ (Simplified using Boolean Algebra, Units 2\&3)}
 \end{aligned}$$

# Threshold Detector (2/2)

- Counting 1's, we have SOP
- What if counting 0's?

	A	B	C	f	f'
0	0	0	0	0	1
1	0	0	1	0	1
2	0	1	0	0	1
3	0	1	1	1	0
4	1	0	0	1	0
5	1	0	1	1	0
6	1	1	0	1	0
7	1	1	1	1	0



Show the condition to make output == 0

0      1      2

$$f' = A'B'C' + A'B'C + A'BC'$$

Taking the complement by DeMorgan's law,

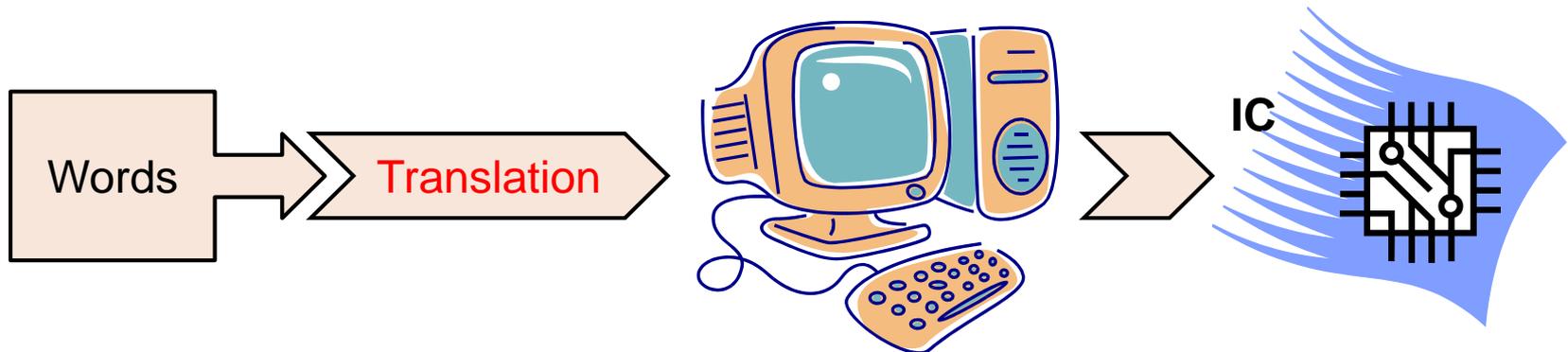
$$\begin{aligned} f &= (A'B'C' + A'B'C + A'BC')' \\ &= [(A' \bullet B' \bullet C') + (A' \bullet B' \bullet C) + (A' \bullet B \bullet C')] \\ &= (A + B + C)(A + B + C')(A + B' + C) \text{ (POS)} \end{aligned}$$

Minterm & maxterm expansions

# Summary: Logic Design using a Truth Table

## □ General steps:

1. Make a **truth table** according to the word description
2. Generate a Boolean expression
  - SOP: check 1's
  - POS: check 0's
    - First of all, have  $f'$  in SOP, then derive  $f$  in POS
3. Simplify the Boolean expression



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# Minterm & Maxterm Expansions

# Minterm vs. Maxterm

- **Definition:** A **minterm/maxterm** of  $n$  variables is a **product/sum** of  $n$  literals in which each variable appears **exactly once** in either true or complement form (**but not both**)
  - ▣ A **literal** is a variable or its complement, e.g.,  $A, A'$
  - ▣ e.g., assume 3 variables
    - Minterm:  $A'BC, AB'C'$
    - Maxterm:  $A+B+C, A+B+C'$

$(m_i)' = M_i$   
DeMorgan

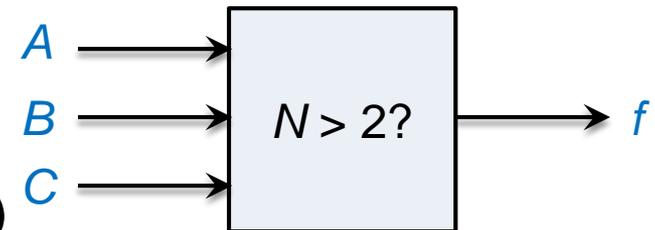
Each minterm corresponding one input condition

Row No.	$ABC$	Minterms $m_i$	Maxterms $M_i$
0	000	$m_0 = A'B'C'$	$M_0 = A + B + C$
1	001	$m_1 = A'B'C$	$M_1 = A + B + C'$
2	010	$m_2 = A'B C'$	$M_2 = A + B' + C$
3	011	$m_3 = A'B C$	$M_3 = A + B' + C'$
4	100	$m_4 = A B'C'$	$M_4 = A' + B + C$
5	101	$m_5 = A B'C$	$M_5 = A' + B + C'$
6	110	$m_6 = A B C'$	$M_6 = A' + B' + C$
7	111	$m_7 = A B C$	$M_7 = A' + B' + C'$

# Minterm Expansion

- A **minterm expansion** or a **standard sum of products**: a function is written as **a sum of minterms**
- $\Rightarrow$  i.e., counting **1's**
- e.g.,

$$\begin{aligned}
 \square f &= A'BC + AB'C' + AB'C + ABC' + ABC \\
 &= m_3 + m_4 + m_5 + m_6 + m_7 \text{ (} m\text{-notation)} \\
 &= \Sigma m(3, 4, 5, 6, 7)
 \end{aligned}$$



Row No.	ABC	Minterms $m_i$	Maxterms $M_i$
0	000	$m_0 = A'B'C'$	$M_0 = A + B + C$
1	001	$m_1 = A'B'C$	$M_1 = A + B + C'$
2	010	$m_2 = A'B C'$	$M_2 = A + B' + C$
3	011	$m_3 = A'B C$	$M_3 = A + B' + C'$
4	100	$m_4 = A B'C'$	$M_4 = A' + B + C$
5	101	$m_5 = A B'C$	$M_5 = A' + B + C'$
6	110	$m_6 = A B C'$	$M_6 = A' + B' + C$
7	111	$m_7 = A B C$	$M_7 = A' + B' + C'$

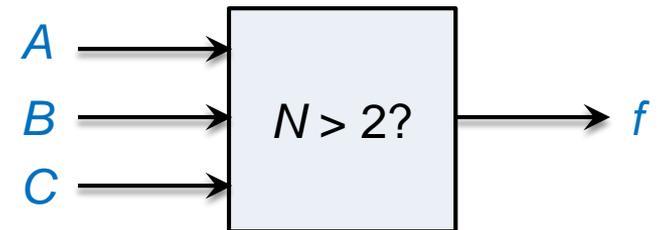
Minterm & maxterm expansions

# Maxterm Expansion

- A **maxterm expansion** or a **standard product of sums**: a function is written as a **product of maxterms**
- $\Rightarrow$  i.e., counting **0's**
- e.g.,

□  $f' = A'B'C' + A'B'C + A'BC'$

$\Rightarrow f = (A + B + C)(A + B + C')(A + B' + C)$   
 $= M_0 M_1 M_2$  (**M-notation**) =  $\Pi M(0, 1, 2)$



Row No.	ABC	Minterms $m_i$	Maxterms $M_i$
0	000	$m_0 = A'B'C'$	$M_0 = A + B + C$
1	001	$m_1 = A'B'C$	$M_1 = A + B + C'$
2	010	$m_2 = A'B C'$	$M_2 = A + B' + C$
3	011	$m_3 = A'B C$	$M_3 = A + B' + C'$
4	100	$m_4 = A B'C'$	$M_4 = A' + B + C$
5	101	$m_5 = A B'C$	$M_5 = A' + B + C'$
6	110	$m_6 = A B C'$	$M_6 = A' + B' + C$
7	111	$m_7 = A B C$	$M_7 = A' + B' + C'$

Minterm & maxterm expansions

# Complement using Minterms/Maxterms

□  $(m_i)' = M_i$

□ **Complement of  $f$ :**

1. Count 0's in  $f$  (find  $f'$  directly)

$$f' = m_0 + m_1 + m_2 = \Sigma m(0, 1, 2)$$

$$f' = M_3M_4M_5M_6M_7 = \Pi M(3, 4, 5, 6, 7)$$

2. Count 1's in  $f$  (find  $f$  and then complement it)

$$f' = (f)' = (m_3 + m_4 + m_5 + m_6 + m_7)'$$

$$= m_3' m_4' m_5' m_6' m_7'$$

$$= M_3 M_4 M_5 M_6 M_7 = \Pi M(3, 4, 5, 6, 7)$$

$$f' = (f)' = (M_0 M_1 M_2)'$$

$$= M_0' + M_1' + M_2'$$

$$= m_0 + m_1 + m_2 = \Sigma m(0, 1, 2)$$

A	B	C	$f$	$f'$
0	0	0	0	1
0	0	1	0	1
0	1	0	0	1
0	1	1	1	0
1	0	0	1	0
1	0	1	1	0
1	1	0	1	0
1	1	1	1	0

# Example: Minterm/Maxterm Expansion

□ e.g.,  $f(a, b, c, d) = a'(b' + d) + acd'$

□  $f(a, b, c, d) = a'(b' + d) + acd'$

$$= a'b' + a'd + acd'$$

$$= a'b'(c + c')(d + d') + a'd(b + b')(c + c') + acd'(b + b')$$

$$= a'b'c'd' + a'b'c'd + a'b'cd' + a'b'cd + a'bc'd + a'bcd + abcd' + ab'cd'$$

$$\begin{array}{cccccccc} 0000 & 0001 & 0010 & 0011 & 0101 & 0111 & 1110 & 1010 \end{array}$$

$$= \Sigma m(0, 1, 2, 3, 5, 7, 10, 14) \text{ (Minterm Expansion)}$$



$$= \Pi M(4, 6, 8, 9, 11, 12, 13, 15) \text{ (Maxterm Expansion)}$$

# Summary (1/2)

- **Convert a Boolean expression to a minterm/maxterm expansion**
  - Use truth table
    - Sometimes there are too many terms
  - Use Boolean algebra
    - SOP → multiply out and use  $(X + X')=1$  → minterm expansion
    - POS → factor and use  $XX'=0$  → maxterm expansion

# Summary (2/2)

□ **Convert between a minterm and a maxterm expansion**

□ If  $f = \Sigma m_j$ , then  $f = \Pi M_j$ , where each  $m_j$  is **not** in  $f$

**DESIRED FORM**

**GIVEN FORM**

	<b>Minterm</b> Expansion of $f$	<b>Maxterm</b> Expansion of $f$	<b>Minterm</b> Expansion of $f'$	<b>Maxterm</b> Expansion of $f'$
<b>Minterm</b> Expansion of $f$	-----	<b>maxterm</b> nos. are those nos. not on the <b>minterm</b> list for $f$	list <b>minterms</b> not present in $f$	<b>maxterm</b> nos. are the same as <b>minterm</b> nos. of $f$
<b>Maxterm</b> Expansion of $f$	<b>minterm</b> nos. are those nos. not on the <b>maxterm</b> list for $f$	-----	<b>minterm</b> nos. are the same as <b>maxterm</b> nos. of $f$	list <b>maxterms</b> not present in $f$

□ **There is a 1-to-1 mapping between a truth table and the minterm/maxterm expansion**

# AND of Two Minterm Expansions

- e.g., Given  $f_1 = \sum m(0, 2, 3, 5, 9, 11)$ ,  $f_2 = \sum m(0, 3, 9, 11, 13, 14)$ , find  $f_1 f_2 = ?$ 
  - AND: take the numbers that appear in both expansions:  
 $f_1 f_2 = \sum m(0, 3, 9, 11)$
  
- Q: What if AND for two maxterm expansions?

# General Truth Table

- **Q: Given three Boolean variables  $A, B, C$ , how many different Boolean functions can you produce?**
- **A:**
  - Each  $a_i$  can be assigned with either 0 or 1
  - $\Rightarrow 2^8 = 256$

$A$	$B$	$C$	$f$
0	0	0	$a_0$
0	0	1	$a_1$
0	1	0	$a_2$
0	1	1	$a_3$
1	0	0	$a_4$
1	0	1	$a_5$
1	1	0	$a_6$
1	1	1	$a_7$

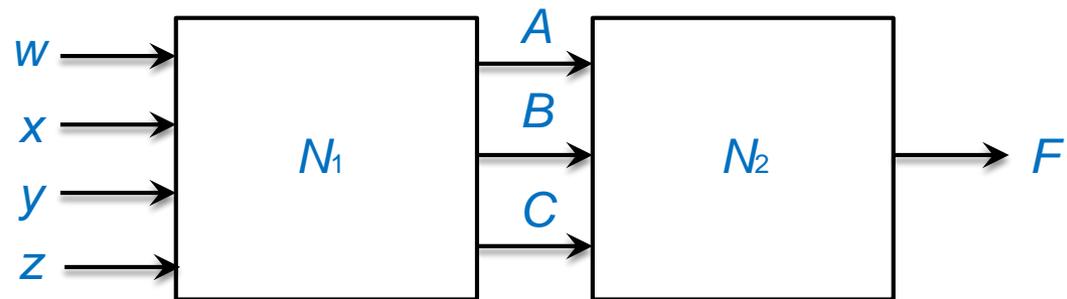
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# Incompletely Specified Functions

# Incompletely Specified Functions (1/2)

- A large digital system is usually divided into subcircuits
  - ▣ Assume  $N_1$  never generates  $ABC = 001/110$  for any  $w, x, y, z$

	A	B	C	F
0	0	0	0	1
1	0	0	1	x
2	0	1	0	0
3	0	1	1	1
4	1	0	0	0
5	1	0	1	0
6	1	1	0	x
7	1	1	1	1



“don't care” (DC) terms can be assigned with either “0” or “1”

- ▣  $F$ : Incompletely specified function
- ▣  $A'B'C, ABC'$ : don't care terms

# Incompletely Specified Functions (2/2)

□ **Impact of don't care terms on Boolean simplification:**

- Try exhaustive combinations of DCs to find the best expression (may be stupid but works for now)

	A	B	C	F
0	0	0	0	1
1	0	0	1	x
2	0	1	0	0
3	0	1	1	1
4	1	0	0	0
5	1	0	1	0
6	1	1	0	x
7	1	1	1	1

1. Assign "0" to both "X":

$$F = A'B'C' + A'BC + ABC = A'B'C' + BC$$

2. Assign "1" to 1<sup>st</sup> "X", and "0" to 2<sup>nd</sup> "X":

$$F = A'B'C' + A'B'C + A'BC + ABC = A'B' + BC$$

3. Assign "1" to both "X":

$$F = A'B'C' + A'B'C + A'BC + ABC' + ABC = A'B' + BC + AB$$

4. ...

- 2. is the simplest solution

□ **Notation:**

- $F = \sum m(0, 3, 7) + \sum d(1, 6)$

- $F = \prod M(2, 4, 5) \bullet \prod D(1, 6)$

← don't care minterm and maxterm have the same numbers

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# Truth Table Construction

# Error Detector for 6-3-1-1 Codes (1/2)

- **Design an error detector for 6-3-1-1 codes:**
  - ▣ The output  $F = 1$  iff inputs ( $A, B, C, D$ ) represent an invalid code combination

Decimal digit	6-3-1-1 code
0	0000
1	0001
2	0011
3	0100
4	0101
5	0111
6	1000
7	1001
8	1011
9	1100

1. Construct the truth table

$ABCD$	$F$
0000	0
0001	0
0010	1
0011	0
0100	0
0101	0
0110	1
0111	0
1000	0
1001	0
1010	1
1011	0
1100	0
1101	1
1110	1
1111	1

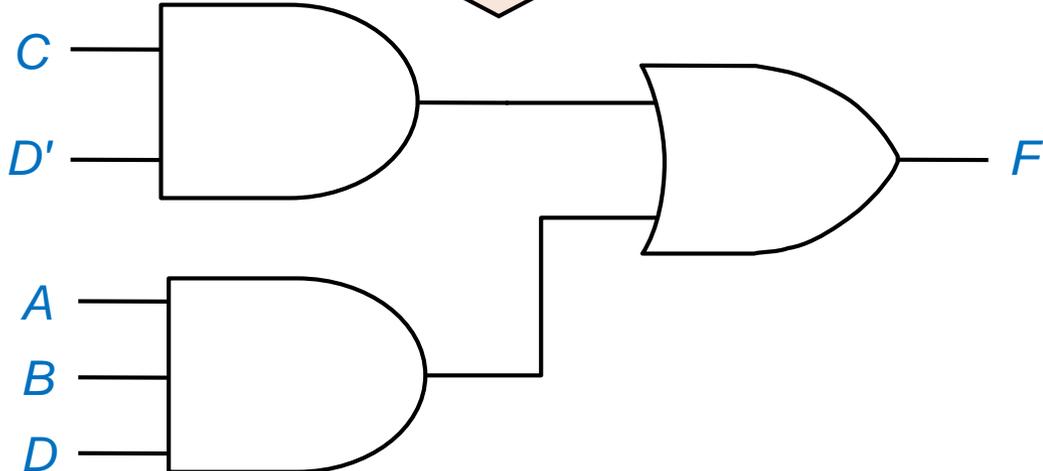
# Error Detector for 6-3-1-1 Codes (2/2)

<i>ABCD</i>	<i>F</i>
0000	0
0001	0
0010	1
0011	0
0100	0
0101	0
0110	1
0111	0
1000	0
1001	0
1010	1
1011	0
1100	0
1101	1
1110	1
1111	1

2. Simplify the function

$$\begin{aligned}
 F(A, B, C, D) &= \Sigma m(2, 6, 10, 13, 14, 15) \\
 &= \underline{A'B'CD'} + \underline{A'BCD'} + \underline{AB'CD'} + \underline{ABCD'} + \underline{ABC'D} + \underline{ABCD} \\
 &= \underline{A'CD'} + \underline{ACD'} + \underline{ABD} \\
 &= CD' + ABD
 \end{aligned}$$

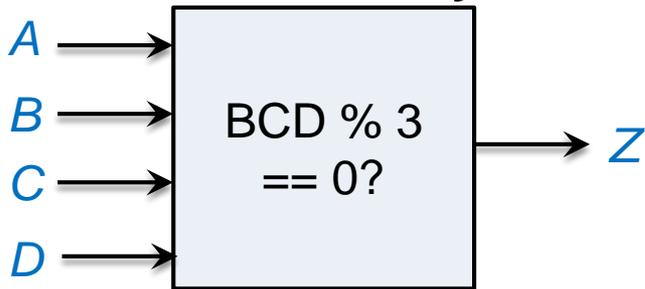
3. Realize it



Minterm & maxterm expansions

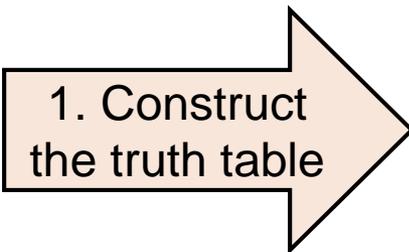
# Example: Truth Table Construction

- The output  $Z = 1$  iff the 8-4-2-1 BCD number ( $A, B, C, D$ ) is divisible by 3

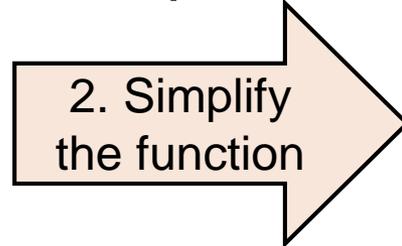


Decimal digit	8-4-2-1 code
0	0000
1	0001
2	0010
3	0011
4	0100
5	0101
6	0110
7	0111
8	1000
9	1001

Minterm & maxterm expansions

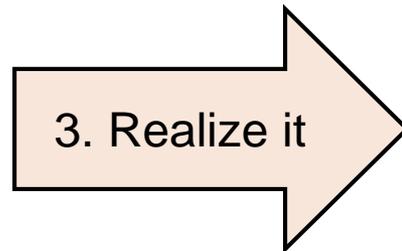


$ABCD$	$Z$
0000	1
0001	0
0010	0
0011	1
0100	0
0101	0
0110	1
0111	0
1000	0
1001	1
1010	X
1011	X
1100	X
1101	X
1110	X
1111	X



$$Z(A, B, C, D) = \sum m(0, 3, 6, 9) + \sum d(10, 11, 12, 13, 14, 15)$$

⇒ Logic minimization using Karnaugh map (Unit 5)



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# Binary Adders and Subtractor

## Divide-and-conquer

# 1-Bit Half Adder (HA)

## □ Addition table:

0	0	1	1
+0	+1	+0	+1
00	01	01	10

carry      sum

1. Construct the truth table

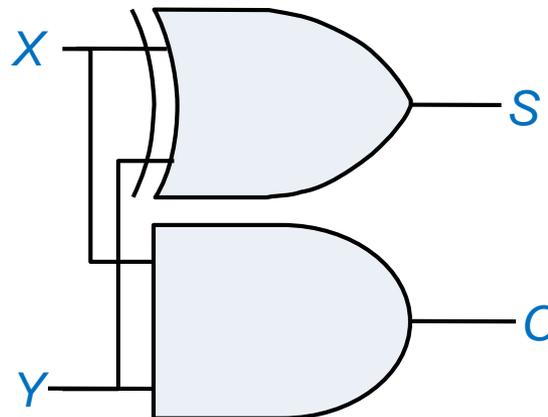
X	Y	Z	C	S
0	0	0	0	0
0	1	0	0	1
1	0	0	0	1
1	1	1	1	0

2. Simplify the function

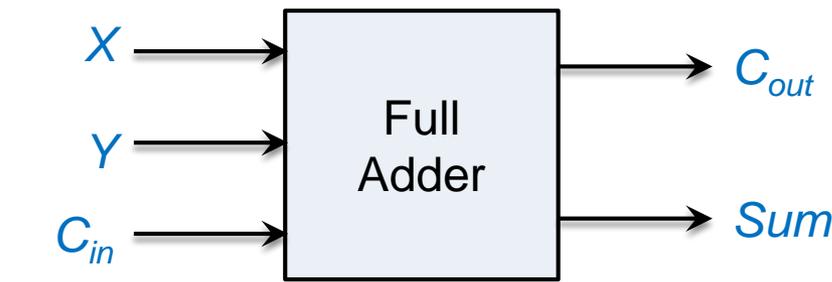
$$C = XY$$

$$S = X'Y + XY' = X \oplus Y$$

3. Realize it



# 1-Bit Full Adder (FA)



2. Simplify the function

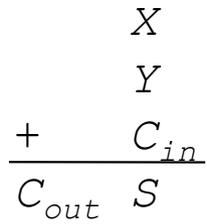
$$C_{out} = X'YC_{in} + XY'C_{in} + XYC_{in}' + XYC_{in}$$

$$= XY + XC_{in} + YC_{in}$$

$$S = X'Y'C_{in} + X'YC_{in}' + XY'C_{in}' + XYC_{in}$$

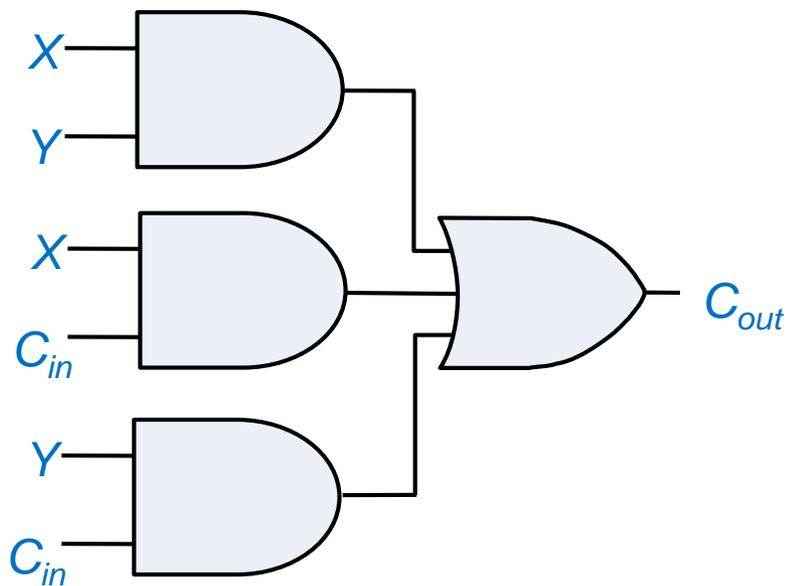
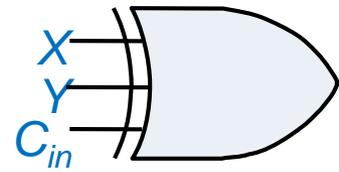
$$= X \oplus Y \oplus C_{in}$$

1. Construct the truth table



$X$	$Y$	$C_{in}$	$C_{out}$	$S$
0	0	0	0	0
0	0	1	0	1
0	1	0	0	1
0	1	1	1	0
1	0	0	0	1
1	0	1	1	0
1	1	0	1	0
1	1	1	1	1

3. Realize it



Minterm & maxterm expansions

# Designing a 4-Bit Parallel Adder (1/3)

□  $\mathbf{A} = (A_3A_2A_1A_0), \mathbf{B} = (B_3B_2B_1B_0)$

$$\begin{array}{r}
 A_3 \ A_2 \ A_1 \ A_0 \\
 + \ B_3 \ B_2 \ B_1 \ B_0 \\
 \hline
 \end{array}
 \quad \longleftrightarrow \quad \mathbf{A} + \mathbf{B}$$

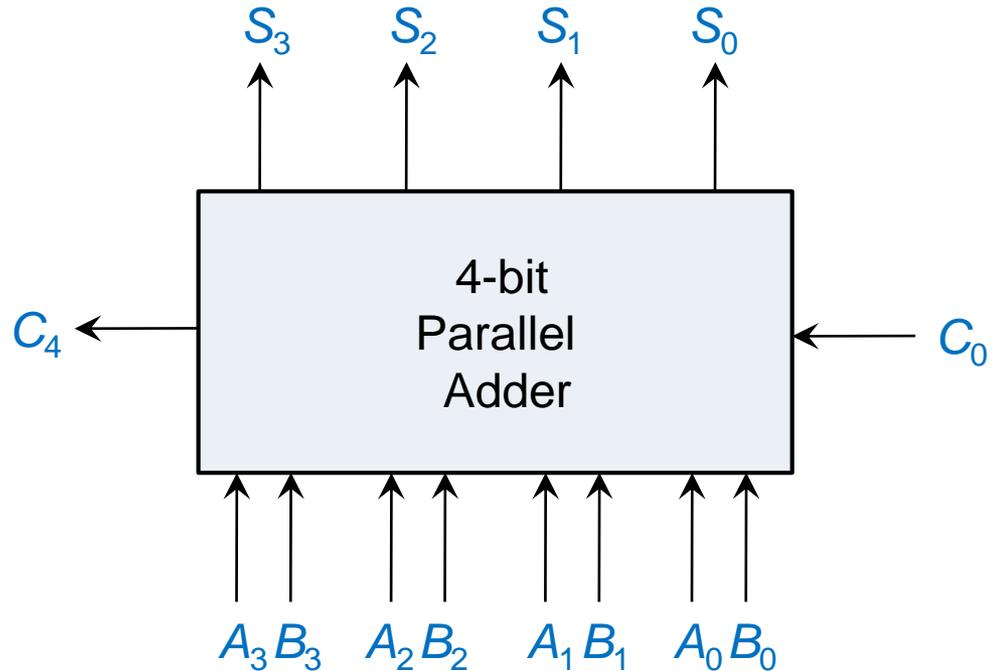
Carry out  $C_4 \leftarrow S_3 \ S_2 \ S_1 \ S_0 \leftarrow C_0$  Carry in

□ e.g.,

$$\begin{array}{r}
 1 \ 0 \ 1 \ 1 \ 0 \ \leftarrow \text{Carries} \\
 1 \ 0 \ 1 \ 1 \\
 1 \ 0 \ 1 \ 1 \\
 \hline
 1 \ 0 \ 1 \ 1 \ 0 \ \leftarrow C_0 (=0)
 \end{array}$$

□ **How?**

- ▣ Use a single truth table?
- ▣  $\Rightarrow$  Too big!



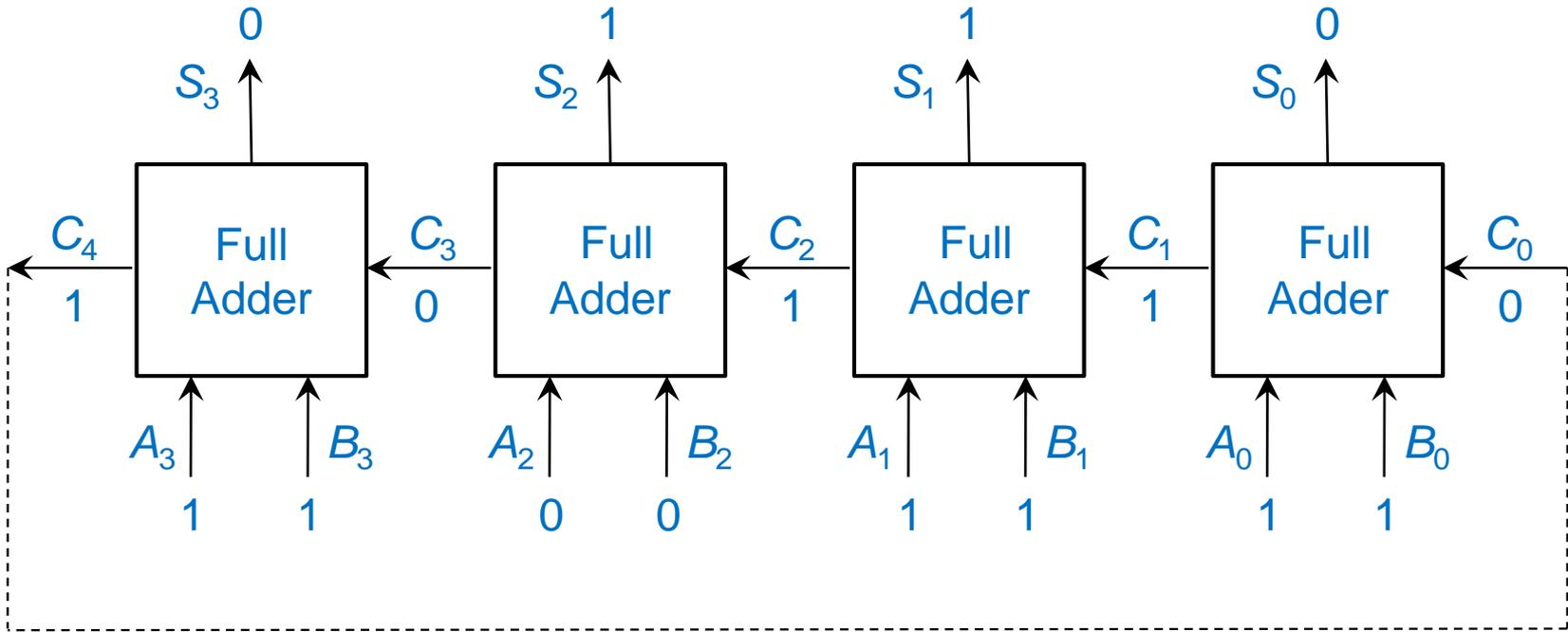
# Designing a 4-Bit Parallel Adder (2/3)

- **Decompose the 4 bit adder into four modules**
  - ▣ Each module adds two bits and a carry  $\Rightarrow$  use full adder
- **Extend to negative numbers**
  - ▣ Consider one's complement (What if two's complement?)
    1. Add just as if all numbers were positive
    2. **End-around carry:** Add the carry out back to the rightmost bit
  - ▣ How to detect overflow?
    - Check **sign!**
    - $(+) + (+) \Rightarrow (-)$  or  $(-) + (-) \Rightarrow (+)$

Add(+A, +B) $A+B < 2^{n-1}$		Add(+A, -B) $B > A$		Add(-A, +B) $B > A$		Add(-A, -B) $A+B \leq 2^{n-1}$	
+3	0011	+5	0101	-5	1010	-3	1100
+4	0100	-6	1001	+6	0110	-4	1011
<u>+7</u>	<u>0111</u>	<u>-1</u>	<u>1110</u>	<u>+1</u>	<u>(1)0000</u>	<u>-7</u>	<u>(1)0111</u>
					 1		 1
					0001		1000

Minterm & maxterm expansions

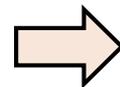
# Designing a 4-Bit Parallel Adder (3/3)



end-around carry for 1's complement

□ **Overflow detection?**

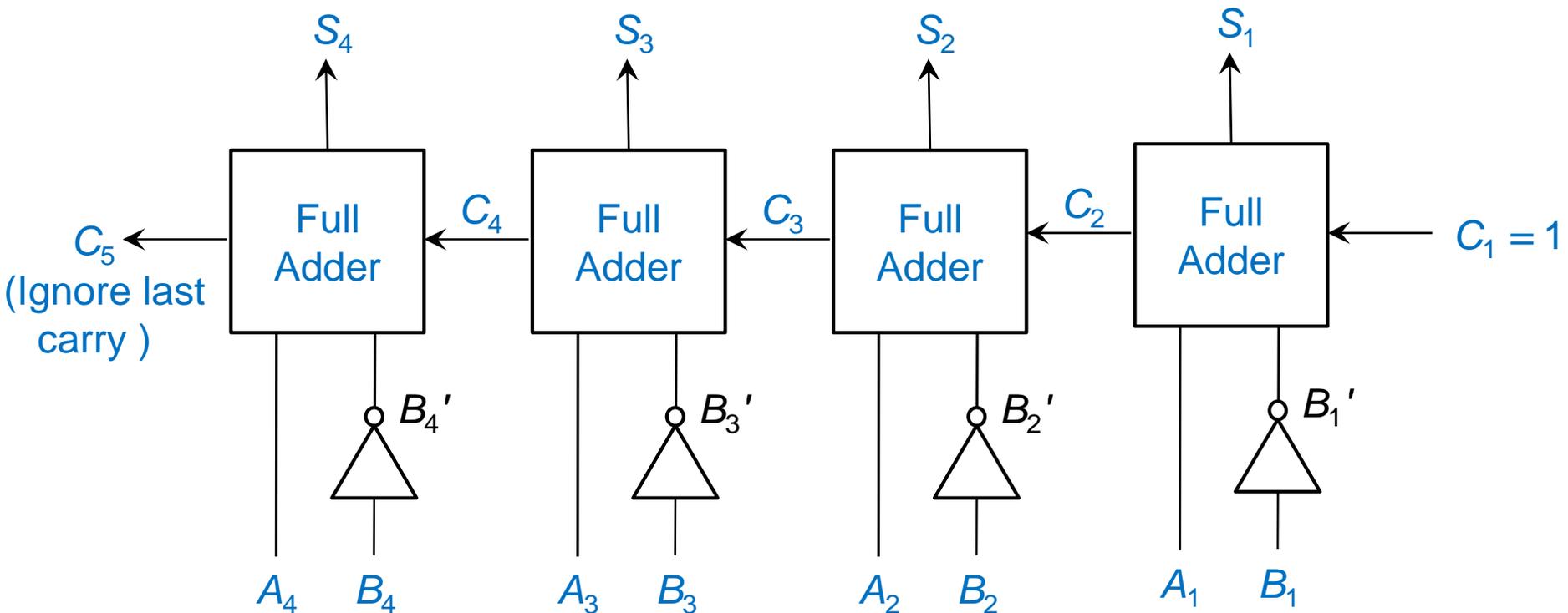
A	B	S	A <sub>3</sub>	B <sub>3</sub>	S <sub>3</sub>	Overflow
+	+	-	0	0	1	1
-	-	+	1	1	0	1



$$V = A_3'B_3'S_3 + A_3B_3S_3'$$

# Designing a Binary Subtractor (1/2)

- Consider  $A - B = A + (-B)$  in 2's complement
  - $A - B = A + (-B) = A + B^* = A + \underline{B} + 1$
  - Convert  $B$  to 2's complement: inverse and then add 1
  - Discard the carry from the sign bit



Minterm & maxterm expansions

# Designing a Binary Subtractor (2/2)

- Or design a full subtractor
  - $X - Y = D$  (difference),  $B$ : borrow
  - DIY!

